

Commissioning and noise study of the ultra-thin chip-on-board PCB for the CALICE SiW-ECAL prototype

S. Callier (OMEGA), Adrián Irles*, R. Poeschl (IJCLab) for the CALICE SiW-ECAL group

*AITANA group at IFIC – CSIC/UV











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CWS2023 adrian.irles@ific.uv.es

Ultra thin PCB: Chip On Board

- ILD tight spatial constrains: Total space for ASICs and PCB 1.8mm (was 1.2mm since ~2007)
 - BGA pcbs + ASICs are ~2.5-3mm





Ultra thin PCB: Chip On Board

Chip-On-Board proposal:

- Naked ASICs wirebonded.
- Cavities (~250um) for the ASICs
- IJCLab & OMEGA collaboration with ITAEC/SKKU (Sungkyunkwan University, Suwon Korea) and EOS company for the PCB production.









COB: chip-on-board



- ~177 Bonding wires
- Bonding by CERN Bondlab
- Regular exchange allowed to iron out early shortcomings

Stephane Callier





- 9+2 layers board
- Overall height ~1.2mm
- ASICs buried in cavities to ensure overall flatness
 - Need to make sure that bonding wires don't pass board surface





More info

FEV ZOO © R. Poeschl

- In recent years the SiW ECAL has developed and used several PCB variants
- To make sure that you don't get lost, here comes an introduction

FEV10-12

FEV_COB

FEV13



- ASICs in BGA Package
- Incremental modifications From v10 -> v12
- Main "Working horses" since 2014



- ASICs wirebonded in cavities
- COB = Chip-On-Board
- Current version FEV11_COB
- Thinner than FEV with BGA
- External connectivity compatible with BGA based FEV10-12



- Also based on BGA packaging
- Different routing than FEV10-12
- Different external connectivity

Current prototype (see later) is equipped with all of these PCBs



▶ In this study, we only consider 6 FEVs (equipped all with 500um)

Pos.	Id	ASU type	An. PS. Dec.	Dig. PS. Dec.
5th	slab 38	FEV13	$(1nF \times 32) \& (10nF \times 32) \& (100nF \times 32) \& (1\mu F \times 32) \& (68\mu F \times 32)$	$(100nF \times 33) \& (33\mu F \times 16)$
6th	slab 39	FEV13	$\begin{split} PA: (1nF \times 32) \& (10nF \times 32) \& (100nF \times 32) \& (1\mu F \times 32) \& (120\mu F \times 32) \\ (1nF \times 32) \& (10nF \times 32) \& (100nF \times 32) \& (1\mu F \times 32) \& (68\mu F \times 32) \\ (PA:)(1nF \times 32) \& (10nF \times 32) \& (100nF \times 32) \& (1\mu F \times 32) \& (120\mu F \times 32) \\ \end{split}$	$(100nF \times 33) \& (33\mu F \times 16)$
7th	slab 29	COB	$(100nF \times 12) \& (150\mu F \times 2) \& (330\mu F \times 5)$	$(100nF \times 14) \& (330\mu F \times 5)$
8th	slab 30	FEV12	$120\mu F \times 16$	$120\mu F \times 16$
9th	slab 33	COB	$(100nF \times 8) \& (150\mu F \times 2) \& (330\mu F \times 5)$	$(100nF \times 8) \& (330\mu F \times 5)$
10th	slab 31	FEV12	$120\mu F \times 16$	$120\mu F \times 16$

Table 1: Summary of devices under study and their main properties. Column by column: Position in the 15 layer stack; ID; ASU type; amount of decoupling capacitors (with their value) for the Analogue Power Supplies and the Digital Power Supplies. Notice that for the FEV13, the preamplifier analogue power supplies (PA) are separated from the other analogue power suplies.











Commissioning: pedestal and noise

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$$\sigma_i^2 = \sigma_{I_i}^2 + \sum_{j=1}^{N_c} \sigma_{C_i^j}^2$$
(1)

The covariance matrix element from the two channels i and k is expressed by:

$$cov(i,k) = \delta_{ik}\sigma_{I_i}\sigma_{I_k} + \sum_{j=1}^{N_c} \sigma_{C_i^j}\sigma_{C_k^j}$$
(2)

where:

$$\delta_{ik} = \begin{cases} 1 & \text{if } i = k \\ 0 & \text{if } i \neq k \end{cases}$$
(3)

The covariance matrix element can also be determined from the data:

$$cov_{Data}(i,k) = \frac{\sum_{n=1}^{N_{event}} (A_i(n) - \mu_{A_i})(A_k(n) - \mu_{A_k})}{N_{event}}$$
(4)
Measured amplitude if
no hit
Adrian.irles@ific.uv.es

Commissioning: pedestal and noise

- For the calculation of the pedestal position for each readout channel and each SCA, the following procedure was followed (at DESY and CERN)
 - 1. Use the default trigger thresholds of ~ 0.5 MIP and the ILC-gain (6pF)
 - 2. Mask all readout channels except channel 0
 - 3. Set-up the injection system to inject 15 signals of between 0.75-1.5 MIP 2 in channel 0. These signals were separated by 100µs each.
 - 4. Make sure that all ASICs from all modules in the stack were recording 15 injected signals.
 - 5. Record 10000 readout cycles of 2ms each.
 - 6. Repeat the process 3 times more but using channels 9, 18 or 27 instead of 0

The analysis is done, chip-wise (treating all chips independently!)





Pedestal and noise, COB -slab29





Pedestal map of COB-slab29 and SCA=8 (high-gain)



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Pedestal and noise, COB -slab29



- For the SCA 0, large variations are observed, specially in the ASICs situated in the second and fourth row.
- Non-convergence due to non-gaussian noise sources

- voltage drops (?) which are translated into pedestal shifts, observed in the data as double pedestal peaks
- Effect enhanced in the sectors with more digital routing lines



Pedestal and noise, COB -slab29



We define a set of "outliers"

- Channels that are at more than 3 sigmas of the average noise (incoherent + coherent in quadrature) of the chip
- Process done iteratively : first we remove the 5sigma outliers for the recalculation of the average
- Removed from the analysis



Outliers per PCB





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7.1	11.00	COD	(100 E + 10) 8 (150 E + 0) 8 (200 E + 5)	(100 E 14) 8 (220 E 5)	
/th	slab 29	COB	$(100nF \times 12) \& (150\mu F \times 2) \& (330\mu F \times 5)$	$(100nF \times 14) \& (330\mu F \times 5)$	
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oui	5140 50	112 12	$120\mu T \times 10$	$120\mu r \times 10$	
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Important remark: before adding the decoupling capacitances to the COBs, the non-gaussian noise source was saturating the readout





Noise (excluding outliers)





Masked channels

Iterative process using 0.5 MIP thresholds as goal and requiring the noise not competing with cosmic signals

FEV12s: (BGA)

- 3% of channels always masked (37 + few channels in near the digital line connectors)
- Less than 1% of random channels

FEV13s (BGA)

- ~1.5% of channels always masked (37 + few channels in near the digital line connectors)
- Less than 1% of random channels

COBs:

- 6-7% of channels systematically masked (not the 37!, all in the digital-lines rows)
- Less than 1% of random channels



MIPs





Id	ASU type	$\langle MIP_{HG} \rangle$ [ADC]	$\langle MIP_{LG} \rangle$ [ADC]
slab 38	FEV13	16.9 ± 1.3	2.7 ± 0.8
slab 39	FEV13	17.3 ± 1.3	2.8 ± 0.8
slab 29	COB	19.0 ± 1.1	3.2 ± 0.9
slab 30	FEV12	17.9 ± 1.0	2.9 ± 0.9
slab 33	COB	18.9 ± 1.6	3.0 ± 0.9
slab 31	FEV12	18.5 ± 1.1	2.9 ± 0.9

Table 3: Summary of devices under study and their average single cell MIP calibration values.



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Summary

▶ Noise levels of the COB are similar to the other solutions.

- Amount of decoupling capacitances seems crucial
- The systematically masked channels is larger

- For all the PCBs, the digital sectors seem critic.
- Next generation of PCBs (FEV2.X) in BGA already existing.
 - Preliminary tests show promising prospects in terms of noise





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Contribution of the coherent noise

$$R_{C1}(chn) = 1 - \frac{\sqrt{\sigma_{I}^{2}(chn) + \sigma_{C2}^{2}(chn)}}{\sqrt{\sigma_{I}^{2}(chn) + \sigma_{C1}^{2}(chn) + \sigma_{C2}^{2}(chn)}}$$
(2)
$$R_{C2}(chn) = 1 - \frac{\sqrt{\sigma_{I}^{2}(chn) + \sigma_{C1}^{2}(chn)}}{\sqrt{\sigma_{I}^{2}(chn) + \sigma_{C1}^{2}(chn)}}$$
(3)



Contribution of the coherent noise

$$R_{C1}(chn) = 1 - \frac{\sqrt{\sigma_{I}^{2}(chn) + \sigma_{C2}^{2}(chn)}}{\sqrt{\sigma_{I}^{2}(chn) + \sigma_{C1}^{2}(chn) + \sigma_{C2}^{2}(chn)}}$$
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(3)

Id	ASU type	$< noise >_{HG} [ADC]$	$< noise >_{LG} [ADC]$	$< R_{C1,HG} > [\%]$	$< R_{C2,HG} > [\%]$	$< R_{C1,LG} > [\%]$	$< R_{C2,LG} > [\%]$
slab 38	FEV13	1.32 ± 0.06	0.74 ± 0.02	4 ± 1	2 ± 1	7 ± 1	$\leq 1 \pm 1$
slab 39	FEV13	1.35 ± 0.07	0.77 ± 0.03	5 ± 1	2 ± 1	9 ± 1	$\leq 1 \pm 1$
slab 29	COB	1.44 ± 0.06	0.74 ± 0.02	5 ± 2	2 ± 1	6 ± 1	$\leq 1 \pm 1$
slab 30	FEV12	1.37 ± 0.06	0.75 ± 0.04	5 ± 1	3 ± 1	8 ± 2	$\leq 1 \pm 2$
slab 33	COB	1.45 ± 0.09	0.74 ± 0.02	6 ± 2	2 ± 1	6 ± 1	$\leq 1 \pm 1$
slab 31	FEV12	1.36 ± 0.08	0.76 ± 0.03	5 ± 2	2 ± 1	8 ± 2	$\leq 1 \pm 1$

Table 2: Summary of devices under study and their noise parameters. Column by column: ID; ASU type; average measured noise values for the *high-gain* and *low-gain* branches; the relative contribution of the coherent noise sources 1 and 2.



